

An Extension of Sandberg's Representation Theorem for Linear Time-Continuous Systems

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Abstract

The input output map of linear time-continuous systems is investigated. The class of all systems for which the input output map is given by the usual integral is investigated. A complete characterization of this class of systems is given in the paper.

Keywords: Linear Systems, Modelling, BIBO-stable Systems, Signal-Processing

1 Introduction

This paper deals with an extension of a representation theorem for linear time-continuous systems recently discovered by I.W. Sandberg. It was shown by I.W. Sandberg, that if a linear time-continuous causal system satisfies certain conditions, then the representation

$$(Sx)(t) = \int_{-\infty}^t h(t, \tau)x(\tau) d\tau + \lim_{a \rightarrow -\infty} (SP_a f)(t) \quad (1)$$

holds for all $t \in \mathbb{R}$. Here the function h has the usual impulse-response interpretation and the function $P_a x$ is given by $(P_a x)(t) = x(t)$ if $t \leq a$ and $(P_a x)(t) = 0$ if $t > a$. So if the signal x is nonzero on a finite time interval $[a, b]$ only then the input-output map of the system S is given for that signal by

$$(Sx)(t) = \int_a^b h(t, \tau)x(\tau) d\tau . \quad (2)$$

The main idea of the theory of time-continuous single-input single-output linear systems is that every such system S has an input-output map that can be represented by

$$(Sx)(t) = \int_{-\infty}^{\infty} h(t, \tau)x(\tau) d\tau . \quad (3)$$

The system S need not be a causal system. Almost always it is emphasized that the representation (3) holds for all time-continuous linear systems [5] [6]. It was recently discovered that such a representation does not hold for all time-continuous single-input single-output linear systems S . The first counter examples were constructed in [3] [8].

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2 Sandberg's Condition

At first we need some preliminaries. Let \mathbb{R} be the set of real numbers, and let $L^\infty(\mathbb{R})$ denote the normed signal space of essentially bounded real valued Lebesgue measurable functions x . The norm is given by

$$\|x\| = \text{esssup}_{t \in \mathbb{R}} |x(t)|. \quad (4)$$

A linear system is a linear map of $L^\infty(\mathbb{R})$ into itself. A linear system S is called causal if the equation $P_a S = P_a S P_a$ holds for all $a \in \mathbb{R}$. We refer to the following two conditions as Sandberg's conditions:

A1) For each t and $a \in \mathbb{R}$ with $a < t$, there is a real constant $c_{t,a}$ such that

$$|(Sx)(t)| \leq c_{t,a} \cdot \int_a^t |x(\tau)| d\tau \quad (5)$$

holds for all $x \in L^\infty(\mathbb{R})$ with $x(\tau) = 0$ for $\tau < a$.

A2) The inequality $\sup_{a \in \mathbb{R}} |(SP_a x)(t)| < \infty$ holds for each $x \in L^\infty(\mathbb{R})$ and $t \in \mathbb{R}$.

Let $\tau \in \mathbb{R}$ be an arbitrary number. We consider for $\delta > 0$ the signal

$$w_{\tau,\delta}(\tau_1) = \begin{cases} \frac{1}{\delta} & \tau_1 \in [\tau, \tau + \delta) \\ 0 & \tau_1 \notin [\tau, \tau + \delta) \end{cases}. \quad (6)$$

The following theorem was shown in [8].

Theorem 1 (Sandberg) *Let S be a causal system that meets Sandberg's conditions A1) and A2). Then the following holds for all $t \in \mathbb{R}$ and all signals $x \in L^\infty(\mathbb{R})$.*

i) *The function*

$$h(t, \tau) = \lim_{\delta \rightarrow 0} (S w_{\tau,\delta})(t) \quad (7)$$

exists for almost all $\tau \in \mathbb{R}$.

ii) *The system S has the representation*

$$(Sx)(t) = \int_{-\infty}^t h(t, \tau) x(\tau) d\tau + \lim_{a \rightarrow -\infty} (SP_a x)(t). \quad (8)$$

Similar results for time-discrete linear systems were obtained in [1] and [2].

At next we want to give an extension of Sandberg's representation theorem. Let p be a positive number which satisfies $1 \leq p < \infty$. We say that a system S satisfies the condition A(p) if the following holds.

For each t and $a \in \mathbb{R}$ with $a < t$, there is a real constant $c_{t,a}(p)$ such that

$$|(Sx)(t)| \leq c_{t,a}(p) \cdot \left(\int_a^t |x(\tau)|^p d\tau \right)^{\frac{1}{p}} \quad (9)$$

holds for all $x \in L^\infty(\mathbb{R})$ with $x(\tau) = 0$ for $\tau < a$.

At first we have to show that the condition A(p) is weaker than the condition A1), that means if the system S satisfies the condition A1) then it also satisfies the condition A(p) for all $p > 1$.

For this let $a < t$ be an arbitrary number and let $x \in L^\infty(\mathbb{R})$ be a signal with $x(\tau) = 0$ for $\tau < a$. Then we have for $p > 1$

$$\begin{aligned} \int_a^t |x(\tau)| d\tau &= \int_a^t |x(\tau)| \cdot 1 \cdot d\tau \\ &\leq \left(\int_a^t |x(\tau)|^p d\tau \right)^{\frac{1}{p}} \cdot \left(\int_a^t 1 d\tau \right)^{\frac{1}{q}} \\ &= (t-a)^{\frac{1}{q}} \cdot \left(\int_a^t |x(\tau)|^p d\tau \right)^{\frac{1}{p}}. \end{aligned} \quad (10)$$

(Here we have used Hölders inequality with $\frac{1}{p} + \frac{1}{q} = 1$.) That means if the system S satisfies the condition A1) with the constant $c_{t,a}$ than it satisfies also the condition A(p) with the constant $(t-a)^{\frac{1}{q}} \cdot c_{t,a}$. In the paper the following result is proved.

Theorem 2 *Let S be an arbitrary causal system such that the conditions A(p) for a $p > 1$ and A2) are met. Then the following holds:*

i) *The function*

$$h(t, \tau) = \lim_{\delta \rightarrow 0} (Sw_{\tau, \delta})(t) \quad (11)$$

exists for almost all $\tau \in \mathbb{R}$ and satisfies

$$\left(\int_a^t |h(t, \tau)|^q d\tau \right)^{\frac{1}{q}} < \infty \quad (12)$$

for each $a \in \mathbb{R}$, where q is given by $\frac{1}{p} + \frac{1}{q} = 1$ [7].

ii) *The input-output map of the system S has the form (8).*

Next, BIBO-stable (bounded input-bounded output) linear systems are investigated. This means that a positive number C_1 exists, such that for all signals $x \in L^\infty(\mathbb{R})$ the inequality

$$\|Sx\| \leq C_1 \cdot \|x\| \quad (13)$$

holds.

Since we have $|(SP_a x)(t)| \leq C_1 \cdot \|P_a x\| \leq C_1 \|x\|$, all BIBO-stable systems satisfy the condition A2). With the Theorem 2 we get the following Corollary.

Corollary 1 *Let S be a causal BIBO-stable linear system such that the condition A(p) is met for some $p \geq 1$. Then the following representation holds for all signals $x \in L^\infty(\mathbb{R})$*

$$(Sx)(t) = (S_1 x)(t) + (S_\infty x)(t). \quad (14)$$

The system S_1 has the form

$$(S_1 x)(t) = \int_{-\infty}^t h(t, \tau) x(\tau) d\tau. \quad (15)$$

The function $h(t, \tau)$ in (15) is given by $h(t, \tau) = \lim_{\delta \rightarrow 0} (Sw_{\tau, \delta})(t)$, which exists for almost all $\tau \in \mathbb{R}$.

The system S_∞ is defined by

$$(S_\infty x)(t) = \lim_{a \rightarrow \infty} (SP_a x)(t) . \quad (16)$$

Both systems S_1 and S_∞ are causal BIBO-stable linear systems.

The limit in (16) exists for all signals $x \in L^\infty(\mathbb{R})$. The representation (14) means that each causal BIBO-stable linear system S which satisfies the condition A(p) can be split into two subsystems S_1 and S_∞ which are given by (15) and (16). The input-output map of the system S_1 is given by the usual integral. The input-output map of the system S_∞ don't has this form. For all signals x which are non zero on a finite time interval only, we have

$$(S_\infty x)(t) = \lim_{a \rightarrow \infty} (SP_a x)(t) = 0 \quad (17)$$

for all $t \in \mathbb{R}$. That means the impulse-response of the system S_∞ is zero for all $t \in \mathbb{R}$. So the system S and S_1 have the same impulse-response.

3 Strongly Continuous Systems

At the end of this paper we investigate those causal BIBO-stable linear systems S , that are characterized by their impulse response. Of course we suppose, that such a system satisfies condition $A(p)$ for some $p \geq 1$. That means to find all causal BIBO-stable linear systems, so that the identity $S = S_1$ holds. For this reason it is necessary to introduce one more concept. A sequence of input signals $x_n \in L^\infty(\mathbb{R})$ is said to be convergent to a signal $x \in L^\infty(\mathbb{R})$, if a positive number C_1 exists, so that

$$\|x_n\| \leq C_1 \quad , n \in \mathbb{N} , \quad (18)$$

and

$$\lim_{n \rightarrow \infty} x_n(t) = x(t) \quad (19)$$

hold for all $t \in \mathbb{R}$.

A causal BIBO-stable linear system S is said to be a strongly continuous system, if for each convergent sequence of signals $x_n \in L^\infty(\mathbb{R})$ the equation

$$(Sx)(t) = \lim_{n \rightarrow \infty} (Sx_n)(t) \quad (20)$$

holds.

Each norm-convergent sequence of signals $x_n \in L^\infty(\mathbb{R})$ is also convergent but not vice-versa. This means the concept of convergence is weaker than the concept of norm-convergence. With these concepts we get the following theorem.

Theorem 3 *Let S be a causal BIBO-stable linear system such that condition $A(p)$ is met for some $p \geq 1$. Then the system S has the form*

$$(Sx)(t) = \int_{-\infty}^t h(t, \tau) x(\tau) d\tau \quad (21)$$

if and only if the system S is strongly continuous.

Theorem (3) characterizes all causal BIBO-stable systems, for which the input-output map is given by the usual integral. Of course the systems have to satisfy condition A(p) for some $p \geq 1$. It can also be shown that the system S in Theorem (3) need not be a causal system. This is a consequence of the following Corollary.

Corollary 2 *Let S be a BIBO-stable strongly continuous linear system such that condition A(p) is met for some $p \geq 1$. Then the system S is also causal.*

4 Proof of Theorem 3

In this section a proof of the Theorem 3 is given.

Let S be an arbitrary causal BIBO-stable linear system such that condition A(p) is met for some $p \geq 1$. Suppose that the system S is strongly continuous. Let $x \in L^\infty(\mathbb{R})$ be an arbitrary input signal. For each $n \in \mathbb{N}$ we consider the signal

$$x_n(t) = \begin{cases} x(t) & |t| \leq n \\ 0 & |t| > n . \end{cases}$$

We have

$$(Sx_n)(t) = \int_{-n}^n h(t, \tau)x(\tau) d\tau . \quad (22)$$

Since the system S is strongly continuous we have

$$\begin{aligned} (Sx)(t) &= \lim_{n \rightarrow \infty} (Sx_n)(t) = \lim_{n \rightarrow \infty} \int_{-n}^n h(t, \tau)x(\tau) d\tau \\ &= \int_{-\infty}^t h(t, \tau)x(\tau) d\tau . \end{aligned} \quad (23)$$

So we have equation (21) for all input signals.

Now let x_n be an arbitrary convergent sequence. Suppose that the input output map of the system S is given by (21). Then we have

$$\begin{aligned} \lim_{n \rightarrow \infty} (Sx_n)(t) &= \lim_{n \rightarrow \infty} \int_{-\infty}^t h(t, \tau)x_n(\tau) d\tau \\ &= \int_{-\infty}^t h(t, \tau)x(\tau) d\tau = (Sx)(t) , \end{aligned} \quad (24)$$

so the system S is strongly continuous. This proves the Theorem 3.

References

- [1] H. Boche, *A Complete Characterization of the Structure of Discrete-Time Linear Systems*, Proc. 15th IMACS World Congress on Scientific Computation, Modelling and Applied Mathematics, Berlin, 1997

- [2] H. Boche and G. Reißig, *Characterization of all Discrete-Time Linear Systems of Convolution Type*, Proc. ECCTD'97, European Conference on Circuit Theory and Design, Budapest, 1997
- [3] S.P. Boyd: *Volterra Series: Engineering Fundamentals*, Phd thesis, Univ. California, Berkeley, 1985
- [4] Th. Kailath, *Modern Signal Processing*, Springer-Verlag, Berlin, 1985
- [5] J.S. Lim, *Fundamentals of Digital Signal Processing*, in [4]
- [6] L. Ljung, *System Identification Theory For The User*, Prentice Hall Information and System Sciences Series, Ed. Th. Kailath, Englewood Cliffs, New Jersey, 1987.
- [7] W. Rudin, *Real and Complex Analysis*, Third Edition, McGraw-Hill Mathematics Series, New York, 1987
- [8] I.W. Sandberg, *A Representation Theorem for Linear Systems*, preprint, Univ. of Texas at Austin, accepted in IEEE CAS I, 1997
- [9] I.W. Sandberg, *Multidimensional Nonlinear Myopic Maps, Volterra Series, and Uniform Neural-Network Approximations*, Proc. of the 4th Workshop on Intelligent Methods for Signal Processing and Communications, Bayona, Spain, June 1996